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Learning to Modulate for Non-coherent MIMO

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This document does not contain Technology as defined in EAR Part 772.





Overview

- · Apply machine learning to design modulation and detection
 - Target: space-time constellations for non-coherent MIMO channel
 - Recent trend: encode/decode with Deep Neural Networks (DNN)
- Simulation-driven, end-to-end, encoder/decoder optimization
 - Minimizing cross-entropy loss \Leftrightarrow maximizing mutual info
 - We compare DNN-based vs DNN-free systems
- Learned schemes can outperform traditional designs at some SNRs
 - DNNs can be avoided altogether while keeping similar performance
 - Feasibility of non-coherent MIMO with only two time slots



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Non-Coherent MIMO Channel

Using $m \operatorname{Tx}$ and $n \operatorname{Rx}$ antennas over L time slots

$\mathbf{Y} = \mathbf{H}\mathbf{X} + \mathbf{Z}$

- Signal $\mathbf{X} \in \mathbb{C}^{m imes L}$ sent over m Tx antennas and L time slots
- Unknown block fading channel $\mathbf{H} \in \mathbb{C}^{n \times m} \stackrel{\mathrm{iid}}{\sim} \mathcal{CN}(0, 1/m)$
- Gaussian noise $\mathbf{Z} \in \mathbb{C}^{n \times L} \stackrel{\text{iid}}{\sim} \mathcal{CN}(0, \sigma^2)$
- Power constraint: $\mathbb{E}[\|\mathbf{X}\|^2/(mL)] = 1$, average SNR $= 1/\sigma^2$
- Receive $\mathbf{Y} \in \mathbb{C}^{n imes L}$ on $n \ \mathsf{Rx}$ antennas

Goal: design k-bit modulation and non-coherent detection scheme

$$\mathbf{m} \in \{0,1\}^k \rightarrow \fbox{\mathsf{ENC}} \rightarrow \mathbf{X} \rightarrow \fbox{\mathsf{MIMO}} \rightarrow \mathbf{Y} \rightarrow \fbox{\mathsf{DEC}} \rightarrow \hat{\mathbf{m}}$$





Encoding Signal Constellation via a Lookup Table

For small k (#bits), lookup table most effective and efficient

- Encoder : $\{0,1\}^k \to \mathbb{C}^{m \times L}$ fully captured by table $\mathbf{C} \in \mathbb{C}^{2^k \times m \times L}$
- Subtract centroid (mean across first axis) to get centered $\overline{\mathbf{C}}$
- Normalize average power of codebook $\widetilde{\mathbf{C}}:=\overline{\mathbf{C}}\sqrt{2^kmL}/\|\overline{\mathbf{C}}\|$
- For message $\mathbf{m} \in \{0,1\}^k$, signal $\mathbf{X}_{\mathbf{m}}$ is selected from $\widetilde{\mathbf{C}}$





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Common alternative: DNN-encoder with one-hot message input

- Vastly over-parameterizes the codebook with extraneous layers
- Binary input encoding would scale better for large k
- Our work uses lookup table to avoid encoder DNN



Decoding with or without DNN

We optimize two soft-decision decoders: DNN-based vs DNN-free

- Both output unnormalized, log-likelihoods for each message
- Apply softmax to yield approximate posterior $P_{\mathbf{m}|\mathbf{Y}}^{\theta}$



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Neural Network (NN) Decoder: θ is network parameters

- Used Multi-Layer Perceptron (MLP) or Residual MLP (ResMLP)
- Blind decoder trained end-to-end with cross-entropy loss



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Pseudo-ML (pML) Decoder: based on orthonormal code ML decoding

- If codeword orthonormal, i.e., $\forall \mathbf{m}, \mathbf{X}_{\mathbf{m}} \mathbf{X}_{\mathbf{m}}^{\dagger} = L \cdot \mathbf{I}_{m}$, then $\|\mathbf{Y}\mathbf{X}_{\mathbf{m}}^{\dagger}\|^{2}$ is proportional to unnormalized, log-likelihood $\log \alpha P(\mathbf{Y}|\mathbf{m})$
- Outputs $\left\{\theta\|\mathbf{Y}\mathbf{X}_{\mathbf{m}}^{\dagger}\|^{2}\right\}_{\mathbf{m}\in\{0,1\}^{k}}$ where $\theta>0$ captures confidence
- Requires additional codebook orthonormality constraint



Cross-Entropy Loss Training Maximizes Mutual Info

$$\mathbf{m} \rightarrow \boxed{\mathsf{ENC}_{\mathbf{C}}} \rightarrow \mathbf{X}_{\mathbf{m}} \rightarrow \boxed{\mathsf{MIMO}} \rightarrow \mathbf{Y} \rightarrow \boxed{\mathsf{DEC}_{\theta}} \rightarrow P_{\mathbf{m}|\mathbf{Y}}^{\theta}$$

End-to-end optimization with cross-entropy loss

$$\min_{\mathbf{C}, \theta} \mathbb{E} \big[-\log P^{\theta}_{\mathbf{m} | \mathbf{Y}}(\mathbf{m} | \mathbf{Y}) \big]$$



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Equivalent to maximizing mutual info $\mathcal{I}(\mathbf{m};\mathbf{Y})$ since

- $\mathbb{E}\left[-\log P_{\mathbf{m}|\mathbf{Y}}^{\theta}(\mathbf{m}|\mathbf{Y})\right] = \mathcal{H}(\mathbf{m}|\mathbf{Y}) + \mathrm{KL}(P_{\mathbf{m}|\mathbf{Y}} \| P_{\mathbf{m}|\mathbf{Y}}^{\theta})$
- $\mathcal{I}(\mathbf{m};\mathbf{Y}) = \mathcal{H}(\mathbf{m}) \mathcal{H}(\mathbf{m}|\mathbf{Y})$, with constant $\mathcal{H}(\mathbf{m}) = k$



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For pML decoder, we also enforce orthonormality with soft penalty

$$\min_{\mathbf{C},\theta} \mathbb{E} \left[-\log P_{\mathbf{m}|\mathbf{Y}}^{\theta}(\mathbf{m}|\mathbf{Y}) \right] \left(1 + \lambda \ell(\mathbf{C}) \right), \quad \lambda > 0$$
$$\ell(\mathbf{C}) := \frac{1}{2^k m^2} \sum_{\mathbf{m} \in \{0,1\}^k} \left\| \mathbf{X}_{\mathbf{m}} \mathbf{X}_{\mathbf{m}}^{\dagger} / L - \mathbf{I}_m \right\|^2$$





Experimental Evaluation

- Non-coherent MIMO parameters
 - Bits $k \in \{2, 4, 6, 8\}$, time slots $L \in \{2, 4\}$
 - **RX** $n \in \{2, 3, 4\}$, TX m = n for L = 4, TX m = 2 for L = 2
- Encoders/decoders optimized over a range of hyperparameters
 - MLP: varied depth and width of fully-connected layers
 - ResMLP: MLP with additional skip connections and batch-norm
 - **pML**: varied λ parameter controlling code orthonormality
- Compare performance vs codes constructed by [Liang, Xia '02]
 - Existing Grassmann code designs require L > m
 - We demonstrate novel feasibility of learning for L = m = 2
- Cross-entropy loss gives approximate lower-bound on throughput:

$$\frac{k - \mathbb{E} \left[-\log P_{\mathbf{m} | \mathbf{Y}}^{\theta}(\mathbf{m} | \mathbf{Y}) \right]}{L} \lessapprox \frac{I(\mathbf{m}; \mathbf{Y})}{L}$$

Compare with capacity lower-bounds of [Yang, Durisi, Riegler '13]



Comparison for L = 4 Time Slots, TX-RX (m, n) = (2, 2)





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NN Decoder Performance for L = 4







NN Decoder Throughput for L = 4





Changes

NN Decoder Performance for L = 2





NN Decoder Throughput for L = 2





Changes

pML Decoder Performance for L = 4





Changes

pML Decoder Throughput for L = 4





Codebook via pML Decoder for k = 2, L = 4, (m, n) = (4, 4)





Summary



- Applied learning to non-coherent MIMO modulation and detection
 - Simulation-driven, end-to-end, space-time constellation optimization
 - Minimizing cross-entropy loss ⇔ maximizing mutual info
 - We offer a DNN-free system as alternative to using DNN
- Learned schemes can outperform traditional designs at some SNRs
 - DNNs can be avoided altogether while keeping similar performance
 - Non-coherent MIMO feasible with only two time slots